

BEE 233 Circuits

Fall 2015

Lab 3: Opamp circuits: Notes on the prelab

6 Pre-lab

6.1 Recording specified opamp parameters for analysis and design

Requires downloading manufacturer's datasheet.

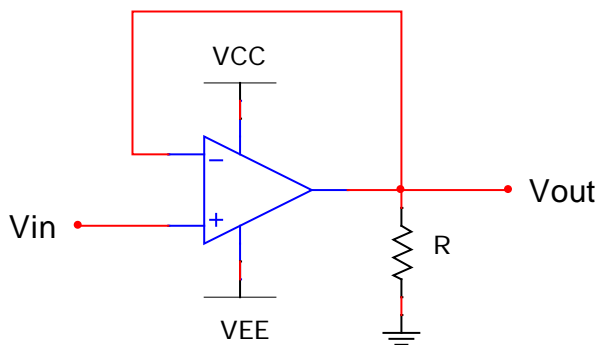
$R_i = 0.8 \text{ M}\Omega$ min to $2.5 \text{ M}\Omega$ typical

$Z_o = 1.0 \text{ }\Omega$ at 1.0 KHz , $A_v = 10.0$

$A_v = 25 \text{ V/mV}$ min to 160 V/mV typical

Slew rate = $0.5 \text{ V}/\mu\text{S}$

6.2 Analysis of simple opamp voltage follower circuit



For the circuit in figure 1 with $R = 5.1 \text{ K}\Omega$, $V_{CC} = +12 \text{ V}$, $V_{EE} = -12 \text{ V}$, and the parameter values in section 6.1, answer the following questions:

1. What is the voltage gain of the circuit at low frequency?

Because we assume the opamp is being operated under conditions where the voltage across the inputs to the opamp will always be zero, this circuit will produce unity gain.

$$V_{in} = V_{out}$$

$$\frac{V_{out}}{V_{in}} = 1$$

2. If V_{in} is a square wave with an amplitude of 2.5 V_{pp} from -1.25 V to $+1.25 \text{ V}$, how long will the output signal take to reach the final value after each input transition?

The slew rate is given as $0.5 \text{ V}/\mu\text{S}$ from the datasheet. For $\Delta V_{in} = 2.5 \text{ V}$ and $A_v = 1$ (unity gain), $\Delta t = 2.5/0.5 = 5 \mu\text{S}$.

3. Assume V_{in} is a sine wave = 3 Vpp. Derive an equation for the maximum rate of change of the output voltage $|dV_{out}/dt|$ as function of the input amplitude and frequency. From this equation and the opamp slew rate, determine the input frequency at which the slew rate of the opamp begins to limit its ability to act as a voltage follower?

$$\text{Let } V_{in} = V_{out} = A \cos(\omega t)$$

$$\frac{dV_{out}}{dt} = -A\omega \sin(\omega t)$$

$$\text{Then } \max \left| \frac{dV_{out}}{dt} \right| = A\omega \text{ occurs where } \sin(\omega t) = -1 \text{ or } +1.$$

$$\text{Since } \frac{dV_{out}}{dt} < \text{Slew rate}, A\omega < 0.5 \text{ V}/\mu\text{S}.$$

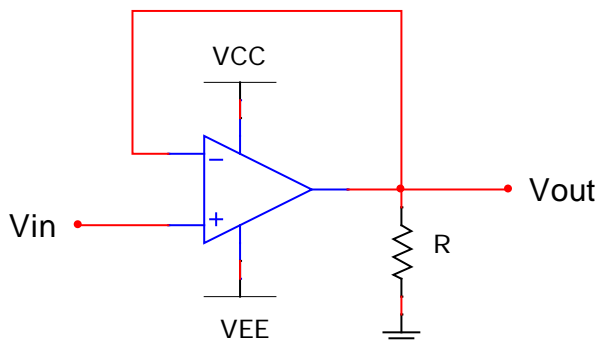
Amplitude is given as 3 Vpp, meaning $A = 1.5 \text{ V}$, therefore:

$$(1.5 \text{ V})\omega < 0.5 \text{ V}/\mu\text{S}$$

$$\omega < 0.333 \text{ radians}/\mu\text{S}$$

$$f = \frac{\omega}{2\pi} = 53 \text{ KHz}$$

4. Assume small-signal inputs to avoid slew-rate limitations and that the opamp is not ideal, i.e., it has a finite open-loop gain A_v . Analyze the circuit in Figure 1 to derive an equation for the circuit gain V_{out}/V_{in} as a function of A_v . At what value of A_v does V_{out}/V_{in} equal 0.5?



$$V_{out} = A_v(V_{in} - V_{out})$$

$$V_{out} = A_v V_{in} - A_v V_{out}$$

$$V_{out}(1 + A_v) = A_v V_{in}$$

$$\frac{V_{out}}{V_{in}} = \frac{Av}{1 + Av}$$

Let $\frac{V_{out}}{V_{in}} = 0.5$. Then

$$\frac{V_{out}}{V_{in}} = \frac{Av}{1 + Av} = 0.5$$

$$0.5 + 0.5 Av = Av$$

$$0.5 = 0.5 Av$$

$$Av = 1$$

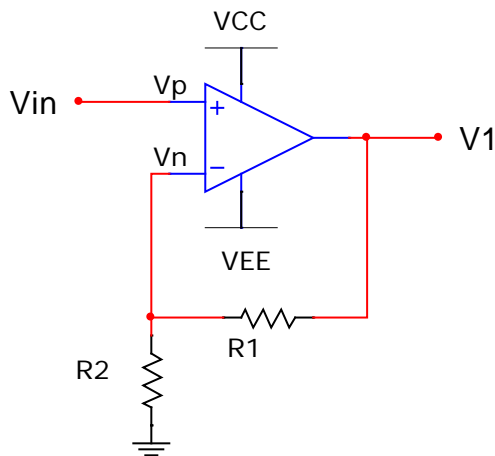
- Using the opamp specifications (plot of opamp gain A_V as function of frequency) and the result in item 4 above, at what frequency do you expect $V_{out}/V_{in} = 0.5$?

From the manufacturers' datasheets, the opamp gain should fall to unity at 1 MHz.

6.3 Analysis of the gain circuit in Figure 2

Use the techniques in the text and what you've learned about opamp circuits, analyze the gain circuit in Figure 2 following this procedure:

- What is the function of the first opamp stage? Find the voltage gain V_1/V_{in} of this stage?



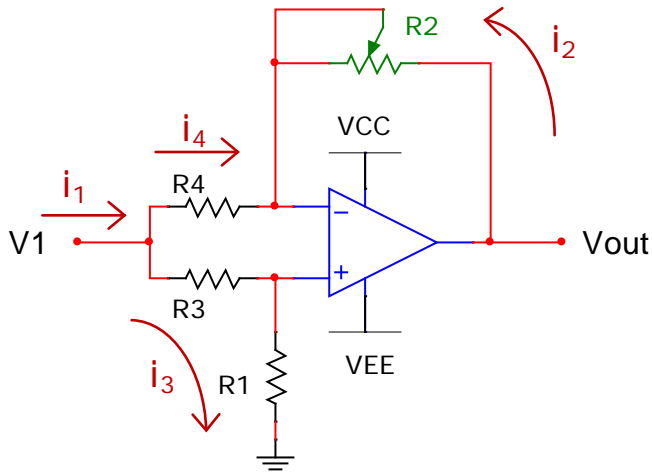
$R_1 = 20 \text{ K}\Omega$, $R_2 = 1 \text{ K}\Omega$. Assume $V_p = V_n$.

$$V_n = V_{in} = \left(\frac{R_2}{R_1 + R_2} \right) V_1$$

$$Av1 = \frac{V1}{Vin} = \frac{R1 + R2}{R2} = \frac{20 + 1}{1} = 21$$

The first stage is a non-inverting amplifier with a gain = 21.

2. What is the function of the second opamp stage? Find the voltage gain $V_{out}/V1$ of this stage as a function of the variable resistance $R2$.



$$i1 = i3 + i4$$

$$i4 = -i2$$

$$V1 = i3(R1 + R3)$$

$$V1 = i4 R4 + Vn$$

$$V1 = i4 R4 + i3 R1$$

From these two equations for $V1$:

$$i3(R1 + R3) = i4 R4 + i3 R1$$

$$i3 R1 + i3 R3 = i4 R4 + i3 R1$$

$$i3 R3 = i4 R4$$

Since $R3 = R4$:

$$i3 = i4$$

$$V1 = i4(R1 + R3)$$

Separately:

$$V_{out} = i_2 R_2 - i_4 R_4 + V_1$$

$$V_{out} = i_2 R_2 - i_4 R_4 + i_4(R_1 + R_3)$$

$$V_{out} = i_4(R_1 - R_2 + R_3 - R_4)$$

Again, since $R_3 = R_4$:

$$V_{out} = i_4(R_1 - R_2)$$

$$A_{v2} = \frac{V_{out}}{V_1} = \frac{i_4(R_1 - R_2)}{i_4(R_1 + R_3)} = \frac{R_1 - R_2}{R_1 + R_3} = 0.5 - \frac{R_2}{10.2 K}$$

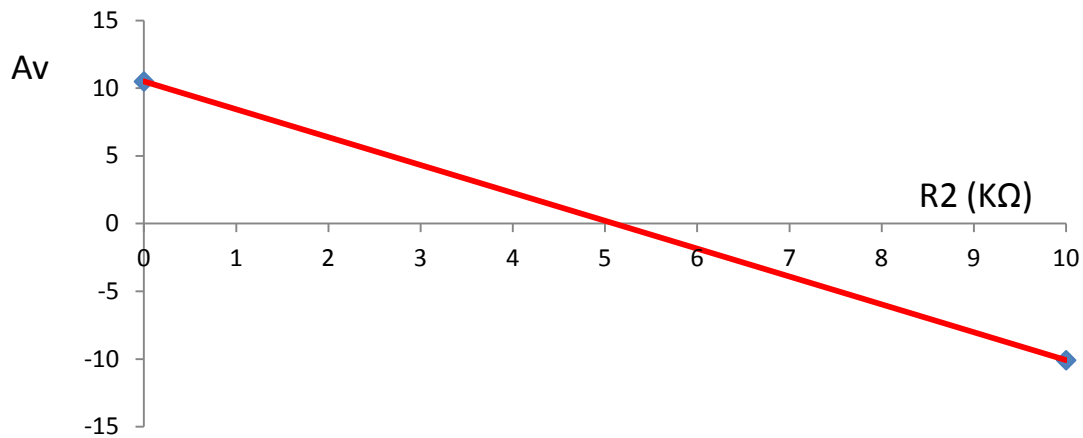
For R_1 and $R_3 = 5.1 K\Omega$ and $0 \leq R_2 \leq 10 K\Omega$:

$$A_{v2} = 0.5 - \frac{R_2}{10.2 K}$$

$$-0.48 \leq A_{v2} \leq +0.5$$

3. With the results from items 1 and 2 above, what is the overall voltage gain V_{out}/V_{in} of this circuit as a function of R_2 ? Plot the gain as function of R_2 . Use the linear scale for the gain.

$$A_v = (A_{v1})(A_{v2}) = (21) \left(0.5 - \frac{R_2}{10.2 K} \right) = 10.5 - \frac{R_2}{485.7}$$



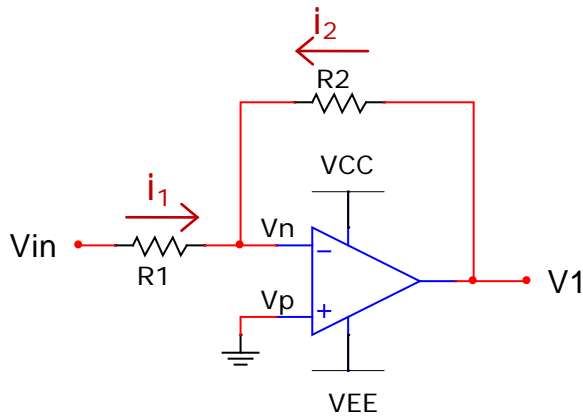
4. You should now be able to see what is interesting about this circuit. Explain its feature in one sentence.

It provides adjustable inverting or non-inverting gain.

6.4 Design of another gain circuit

Re-design the circuit in Figure 2 so that the new overall gain has the opposite sign, i.e. if the circuit in Figure 2 has a gain G , the new circuit has gain $-G$ over the entire range of the resistor $R2$. Use as few components as possible and keep the design simple.

The easiest solution is to replace the non-inverting first stage with an inverting first stage of the same gain.



$$V_n = V_p = 0$$

$$i_1 = -i_2$$

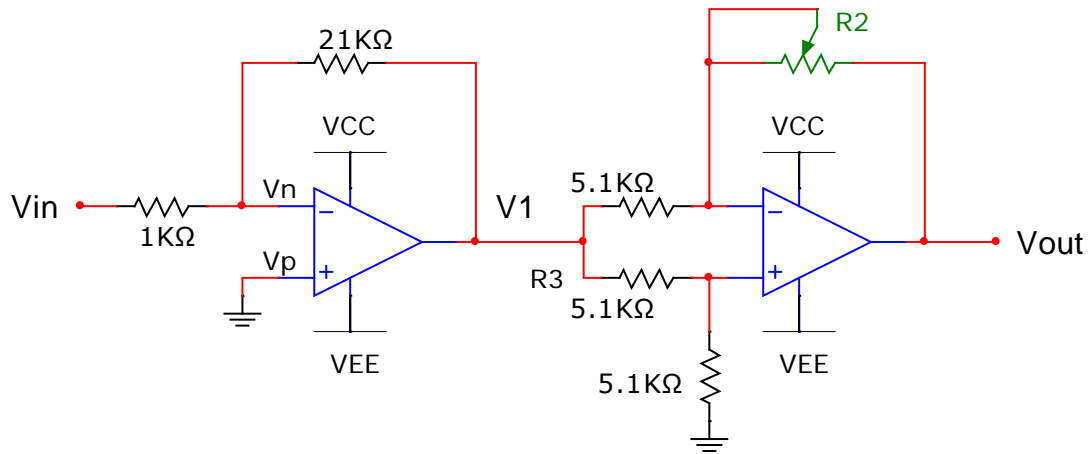
$$V_{in} = i_1 R_1 + V_n = i_1 R_1$$

$$V_1 = i_2 R_2 + V_n = i_2 R_2 = -i_1 R_2$$

$$A_{v1} = \frac{V_1}{V_{in}} = -\frac{i_1 R_2}{i_1 R_1} = -\frac{R_2}{R_1}$$

$$\text{For } A_{v1} = -21 \text{ and } R_1 = 1 \text{ K}\Omega, R_2 = 21 \text{ K}\Omega$$

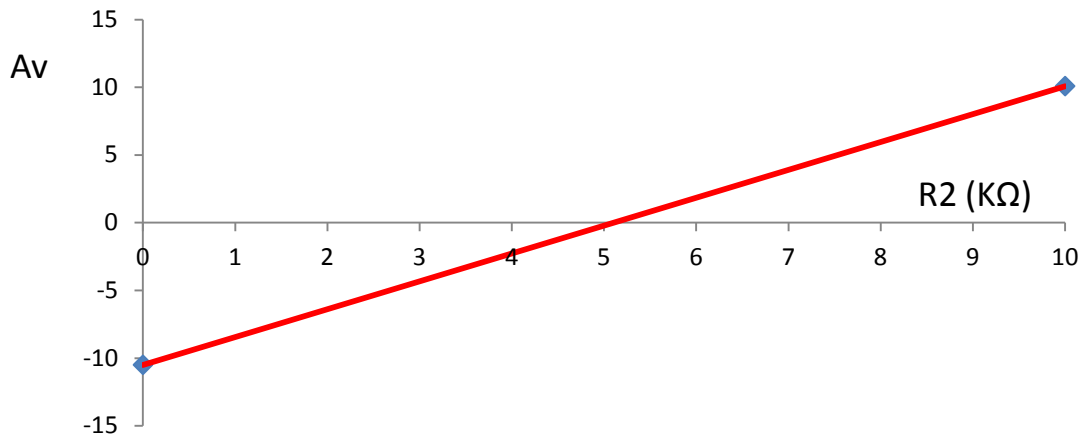
1. Show the schematic of your circuit with all components completely specified (component types and values, component part numbers, power supply values, etc.).



2. Analyze your circuit to prove that it has the gain as specified. If you find out that the magnitude of the gain somehow is not large as the gain magnitude for the circuit in Figure 2, explain why this is so.

$$A_v = (A_{v1})(A_{v2}) = (-21) \left(0.5 - \frac{R_2}{10.2 \text{ K}} \right) = -10.5 + \frac{R_2}{485.7}$$

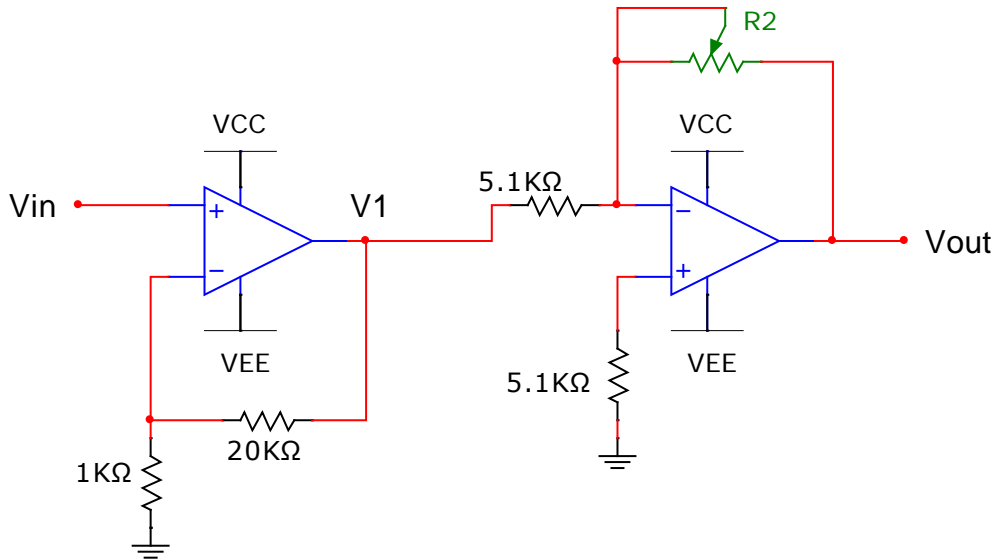
3. Plot the gain of this new circuit as a function of R2. Use the linear scale for the gain.



6.5 Open fault in circuit in Figure 2

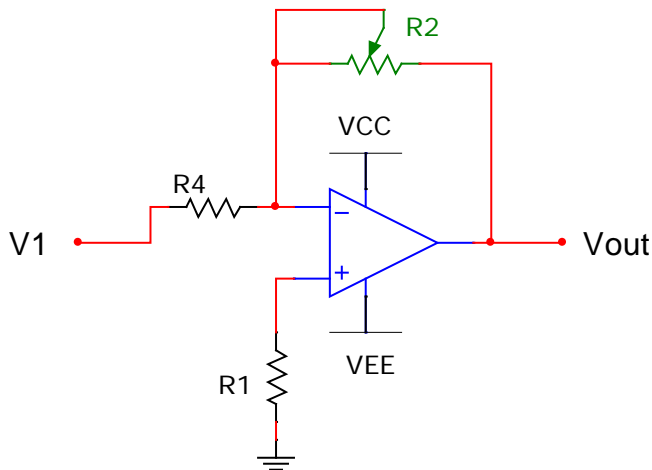
Assume that the circuit in Figure 2 has an open fault at the R3 (5.1 KΩ) resistor. The effect of this open fault is to remove R3 totally from the circuit.

1. Re-draw the circuit diagram in Figure 2, omitting the resistor R3 to simulate the effect of the open fault.



2. Analyze this new circuit to find the overall voltage gain V_{out}/V_{in} in one particular case when $R_2 = 8\text{ K}\Omega$.

Because no current flows through R_1 , the second stage is an inverting amp like the one analyzed in 6.4.



$$A_{v2} = -\frac{R_2}{R_4} = -\frac{8\text{K}}{5.1\text{K}} = -1.57$$

Because the first stage is unchanged, the overall gain is:

$$A_v = (A_{v1})(A_{v2}) = (21)(-1.57) = -32.9$$

3. Is this gain different than the gain when the circuit has no fault? The good (no fault) circuit is also called the "fault-free" circuit.

Yes, it's different. For the good circuit with $R_2 = 8\text{ K}\Omega$:

$$Av_2 = 0.5 - \frac{R_2}{10.2 K} = 0.5 - \frac{8 K}{10.2 K} = -0.284$$

$$Av = (Av_1)(Av_2) = (21)(-0.284) = -5.97$$

Quite clearly, Av is different, -5.97 in the good circuit versus -32.9 in the faulty circuit.